# **%KENWOOD®**

## **ELECTRONIC CROSSOVER**

7 LED ASSY (B30-0451-08)

2 KNOB, PUSH
(K27-0928-08)

SLOPE SLOPE LEVEL

STUDING
TUNING

SYSTEM

SUBSCINIC

SYSTEM

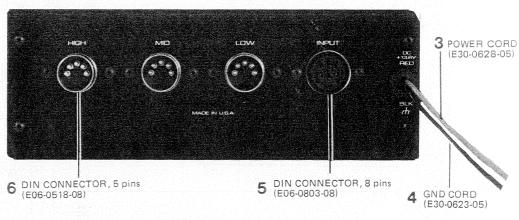
SUBSCINIC

SYSTEM

SUBSCINIC

S

1 KNOB, ROTARY (K29-0924-04)



## **SPECIFICATIONS**

INPUT signal level maximum 1V
Impedance
OUTPUT signal level adjustable 0-3V (+10dB)
Impedance 150 to 275 ohm
RESPONSE 10Hz to 100kHz @ +0/-1.0dB
DISTORTION 0.006% @ 1kHz/1V
THD 0.04% @ 20Hz — 20kHz, 1V output
S/N
SUBSONIC filter Frequency = 20Hz (-3dB)
6dB/octave

	(continuously adjustable)
LO	50Hz to 1.6kHz
	50Hz to 1.6kHz
MID HI	1.6kHz to 8.0kHz
HI	1.6kHz to 8.0kHz
CROSSOVER SLOPE	(switchable)
Filters	6dB/octave, 12dB/octave
POWER	+13.8V, 75mA
DIMENSIONS	(W) 5¾ (H) 2 (D) 3½ inches
WEIGHT	2 lb



### CIRCUIT DESCRIPTION

The KEC 1000 is a U.S. designed and manufactured Active Electronic Crossover for automotive use, executed to Professional Audio Standards.

(For component designations, refer to schematic Page 4.)

### **POWER CIRCUIT**

The 13.8V (nominal) DC input is coupled through an inline fuse for circuit protection while diode D1 provides protection against damage from power connection reversal.

DC filtration is achieved by inductor L1 and capacitor C1, forming a passive second order low pass filter. A series pass Darlington transistor Q1 is used as power switch, turning on when 12V DC is applied from input connector pin 3, originated by program source equipment. This CONTROL voltage is coupled through a first order low pass filter (R2 and C2) to the base of Q1, here establishing a voltage ramp at the emitter. Upon removal of the control voltage, R1 provides a discharge path for C2.

Capacitor C3 adds DC filtration and energy storage while C4 and C5 contribute localized high frequency power decoupling.

A buffered reference ( $^{
m V}_{
m R}$ ) is derived from the filtered DC power by a voltage divider (R3 and R4) with C6 providing HF decoupling. OpAmp U7 is connected as a unity gain voltage follower for  $^{
m V}_{
m R}$ .

### SYSTEM SELECT

Considerations when system selection is made:

- A. The position of the System Select switch;
- B. The combination of output connectors used. (The MID-RANGE OUTPUT is always used.)

FIG 1: OUTPUT SELECTION

0)/07514	OUTPUTS			
SYSTEM	LOW	MID	HIGH	
SUB	0	0	_	
2-A	0	0	_	
2-8		0	0	
3	0	0	0	

CIRCLES INDICATE OUTPUT USED IN RESPECTIVE SYSTEM MODES.

In the SUB (woofer) mode the two MID filters are bypassed leaving the MID signal unaffected. The LO signal chain feeds the woofers.

FIG II: ACTIVE FILTERS

EU TED	SYSTEM			
FILTER	SUB	2A	28	3
LOW	0	0	-	0
MID LOW	_	0		0
MID HIGH		_	0	0
HIGH	_	_	0	0

CIRCLES INDICATE FILTERS USED IN RESPECTIVE SYSTEM MODES.

In 2-WAY A operation, the MID LO filter establishes a crossover match with the LO RANGE filter in the 50 Hz to 1.6 kHz tuning range.

2-WAY B selection substitutes the MID HIGH filter, used with the HI RANGE filter and its matching 1.6kHz to 8kHz tuning range.

In the 3-WAY mode both MID filters are used to establish the midrange bandpass of the two crossover points.

#### SIGNAL PATH AND FILTERS

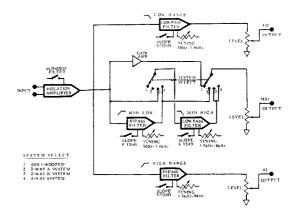
The audio signal is input via connector J1; pin 1 Left, pin 4 Right, and pin 2 Common.

The differential OpAmp U1 provides rejection of unwanted noise which may appear on the input signal lines. The amplifier provides a 4dB voltage gain and combines with a subsonic filter consisting of capacitor C100/C200. When switched into circuit this yields a first order, -6dB/octave highpass filter with -3dB response at 20Hz.

The output of U1 is then split five ways: (see Block Diagram Fig. III).

- 1. To LOW RANGE filter
- 2. To GAIN STAGE
- 3. To SYSTEM SELECT switch
- 4. To MID LOW filter
- 5. To HI RANGE filter

FIG III: BLOCK DIAGRAM



With the SLOPE push switches OUT, the filters are first order (-6dB/octave) tunable frequency filters and are followed by IC OpAmps having a 6dB voltage gain. With switches IN, they become Voltage Controlled Voltage Source (VCVS) second order (-12dB/octave) tunable frequency filters, each with a 6dB voltage gain.

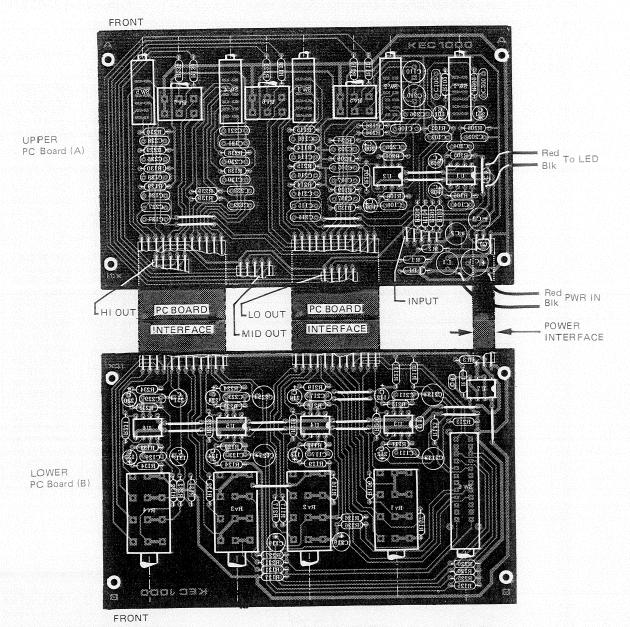
Low pass filters (LOW RANGE and MID HI) have a damping ratio of 1.0 to insure low overshoot in transient response. The high pass filters (MID LO and HI RANGE) have a damping ratio of 0.707 (Butterworth) for flatness of response.

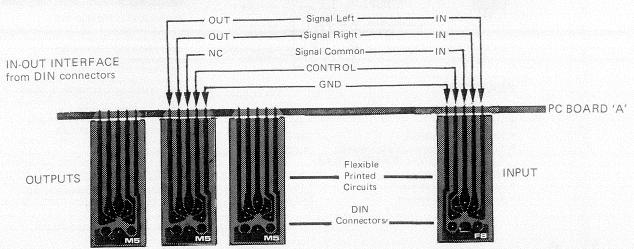
All outputs are capacitively coupled into output level control potentiometers, preceeding respective output connectors.

## **KEC-1000**

### PC BOARDS

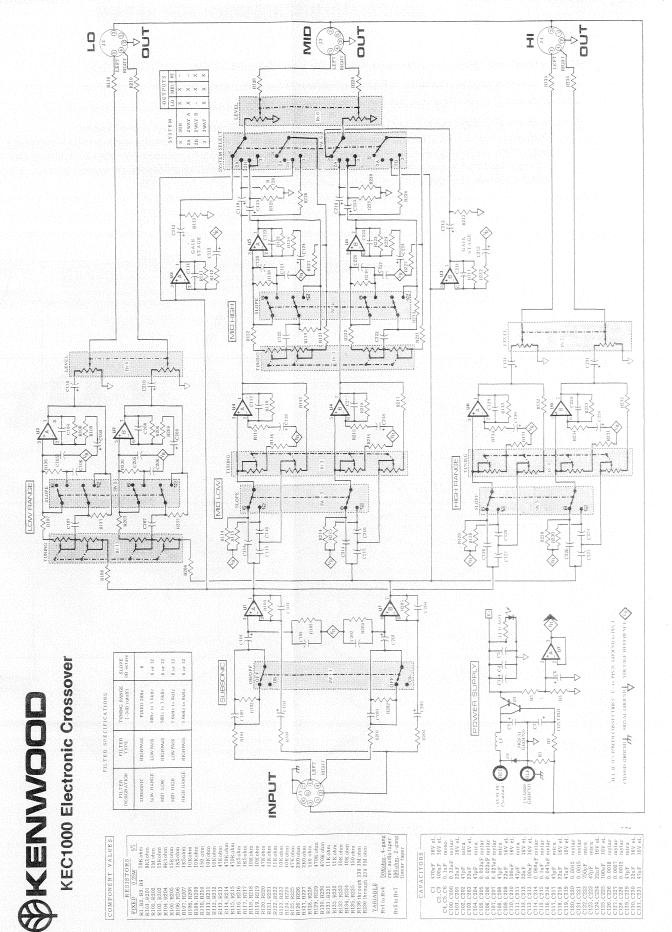
Foil side view. Boards are shown spread apart, joined by interface conductor assemblies.





## SCHEMATIC







### DISASSEMBLY FOR REPAIR

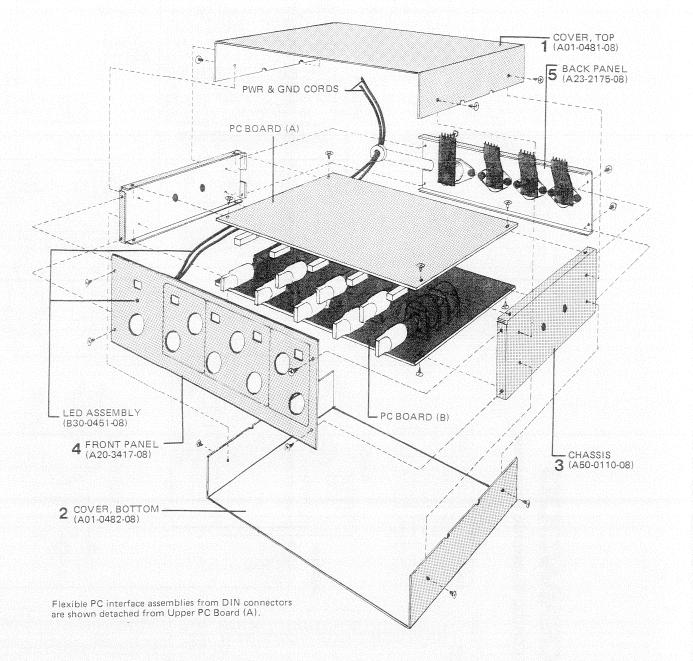
The exploded view below shows the main mechanical parts of the KEC1000.

In most service situations, however, only partial disassembly will be required.

FOR PC BOARD FOIL SIDE ACCESS, remove screws on sides of covers and slip off.

FOR PC BOARD COMPONENT SIDE ACCESS, pull off lower row of knobs.

The reafter remove LOWER PC BOARD (B) holding screws only and fold front edge of board away from assembly. The board will remain functionally attached to Board (A) via board interface assemblies along rear edge of both PC boards permitting powered testing and service, without further disassembly.





## **PARTS LIST**

Ref. No. page 1, photos, front & back panels

Ref. No. page 4, schematic

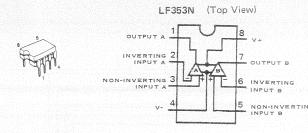
Ref. No. page 5, disassembly drawing

REF. NO.		PARTS NO.	DESCRIPTION
PAGE		TAITION.	DESCRIPTION
5 5 5 5 5 1 1	1 2 3 4 5 - 1 2 3	A01-0481-08 A01-0482-08 A50-0110-08 A20-3417-08 A23-2175-08 J21-3218-08 K29-0924-04 K27-0928-08 E30-0628-05	COVER, TOP COVER, BOTTOM CHASSIS FRONT PANEL BACK PANEL MOUNTING BRACKET KNOB, ROTARY KNOB, PUSH POWER CORD
1		E30-0623-05 B-50-4631-08 B51-1331-08 H21-0217-08	GND CORD INSTRUCTION MANUAL SERVICE MANUAL UNIT CARTON
ELECTRONIC PARTS Ref. page 4			
U1-U Q D' LED L1 Rv1-F Rv5-F SW1-{ J1 J2-J4	1 1 1 Rv4 Rv7 V6 SW5	LF-353-N MPS-A13 1N4003 B30-0451-08 L33-0292-08 R11-4025-08 R06-0002-08 S29-4015-08 S40-4049-08 E06-0803-08 E06-0518-08	DUAL AMP(IC) TRANSISTOR DIODE LED ASSY CHOKE COIL VAR. RESISTOR 50K $\Omega \times 4$ VAR. RESISTOR 500 $\Omega \times 2$ SWITCH, ROTARY, 4W4P SWITCH, PUSH, 2W4P DIN CONNECTOR, 8P DIN CONNECTOR, 5P

Parts without PARTS NO. are not supplied.

REF. NO.	PARTS NO.	DESCRIPTION
C1 C2, C3 C4, C5, C6, C100, C200 C101, C201 C102, C202 C103, C203 C104, C204 C105, C205 C106, C206 C107, C207 C108, C208 C109, C209 C110, C210 C111, C211 C112, C212 C113, C213 C114, C214 C115, C215 C116, C216 C117, C217 C118, C218 C119, C219 C120, C220 C121, C221 C122, C222 C123, C223 C124, C224 C125, C225 C126, C226 C127, C227 C128, C228 C129, C229	C90-0820-05 C24-1210-77 C90-0824-05 C91-0114-05 C90-0825-05 C71-1747-05 C90-0825-05 C71-1747-05 C45-1733-36 C45-1722-36 C45-1727-35 C71-1747-05 C90-0825-05 C24-1210-77 C45-1768-36 C45-1747-05 C90-0825-05 C24-1210-77 C45-1747-05 C90-0825-05 C91-0115-05 C90-0825-05 C91-0115-05 C91-0115-05 C91-0115-05 C90-0825-05 C91-0115-05 C91-0115-05 C91-0115-05 C91-0115-05 C48-1775-15 C90-0825-05 C48-1775-15 C71-1747-05 C91-0825-05 C48-1775-15 C71-1747-05 C48-1775-15 C71-1747-05	CAP. 470uf 16V el CAP. 100uf 16V el CAP. 0.1uf monolithic CAP. 0.33uf mylar CAP. 22uf 16V el CAP. 47pf mica CAP. 22uf 16V el CAP. 47pf mica CAP. 0.033uf mylar CAP. 0.033uf mylar CAP. 0.022uf mylar CAP. 0.022uf mylar CAP. 0.027uf mylar CAP. 100uf 16V el CAP. 100uf 16V el CAP. 100uf 16V el CAP. 100uf 16V el CAP. 0.068uf mylar CAP. 0.047 mylar CAP. 0.047 mylar CAP. 0.015uf mylar CAP. 22uf 16V el CAP. 100uf 16V el CAP. 0.015uf mylar CAP. 22uf 16V el CAP. 100uf 16V el CAP. 100uf 16V el CAP. 100uf 16V el CAP. 100uf 16V el CAP. 22uf 16V el CAP. 100uf 16V el CAP. 100uf 16V el CAP. 100uf 16V el CAP. 0.0018uf mylar CAP. 750pf mica CAP. 47pf mica CAP. 22uf 16V el CAP. 0.0018uf mylar CAP. 0.0018uf mylar CAP. 0.0018uf mylar CAP. 0.0012uf mylar CAP. 0.0012uf mylar CAP. 0.0012uf mylar CAP. 0.0012uf mylar CAP. 47pf mica
C130, C230 C131, C231	C24-0825-05 C90-0822-05	CAP. 22uf 16V el CAP. 47uf 16V el

### **Operational Amplifiers**



### **FEATURES**

8	Wide gain bandwidth	4 MHz
20	High slew rate	13 V/us
22	High input impedance	1012 Ω
	Low total harmonic distortion Av=10,	< 0.02%
	R <sub>L</sub> =10k, V <sub>O</sub> =20 Vp-p, BW=20 Hz-20 kH	-lz
	I averting the later account	

 Low input bias current 50 pA Low input noise voltage 16 nV/√Hz Low input noise current 0.01 pA/√Hz

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ITX 1982 PRINTED IN U.S.A.

B51-1331-08